

Signals and Systems

#08: Multiplication property

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Some of the images in this presentation are from “Signals and Systems”, A. V. Oppenheim, A. S. Wilsky, and S. H. Nawab, 2nd ed. Pearson.

Outline

- #01 Continuous-time signals
- #02 Continuous-time systems
- #03 Fourier series of periodic signals
- #04 Fourier transform of aperiodic signals
- #05 Fourier transform of periodic signals
- #06 Basic properties of the Fourier transform
- #07 Convolution property
- **#08 Multiplication property**
- #09 First-order systems

Multiplication property

We saw that

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The converse holds

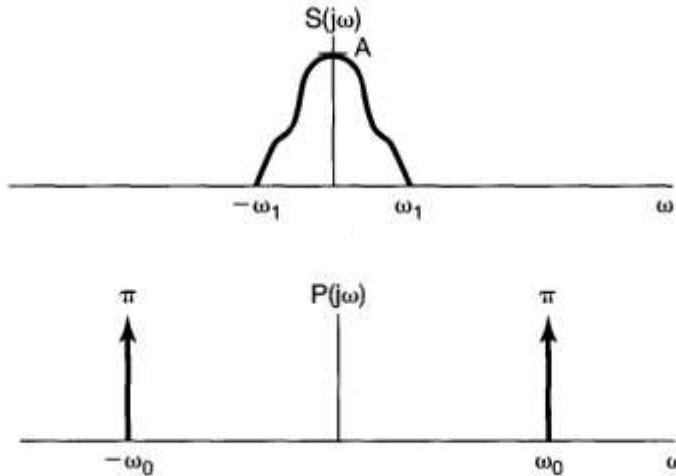
$$r(t) = s(t) \cdot p(t) \iff R(j\omega) = \frac{1}{2\pi} S(j\omega) * P(j\omega) \triangleq \frac{1}{2\pi} \int_{-\infty}^{\infty} S(j\theta) P(j(\omega - \theta)) d\theta$$

Examples

Example: modulation

$$p(t) = \cos \omega_0 t$$

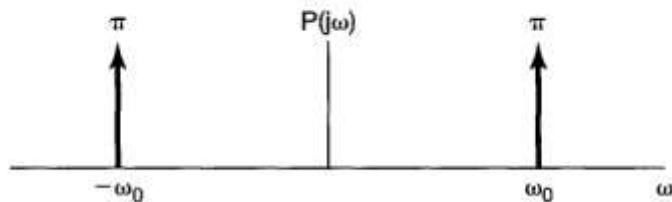
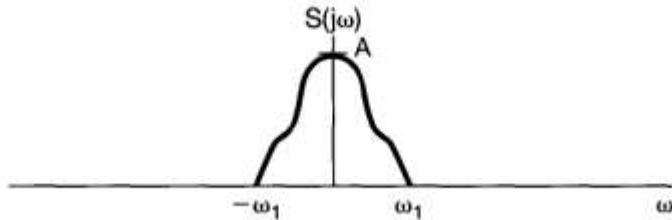
$$P(j\omega) = \pi\delta(\omega - \omega_0) + \pi\delta(\omega + \omega_0)$$



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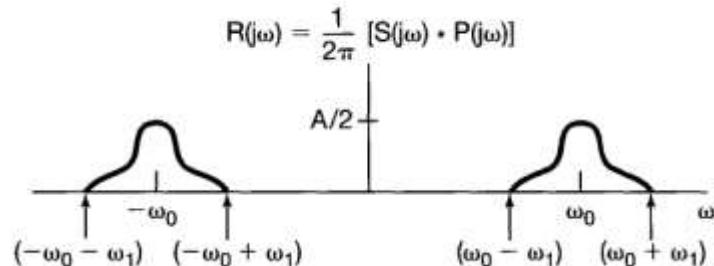
$$P(j\omega) = \pi\delta(\omega - \omega_0) + \pi\delta(\omega + \omega_0)$$



$$r(t) = s(t) \cdot p(t)$$

$$R(j\omega) = \frac{1}{2\pi} S(j\omega) * P(j\omega)$$

$$= \frac{1}{2} S(j(\omega - \omega_0)) + \frac{1}{2} S(j(\omega + \omega_0))$$

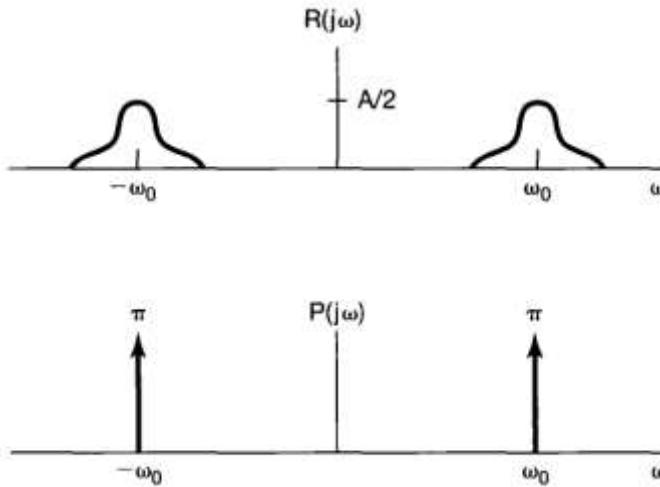


information preserved but shifted at higher frequencies ($\omega_0 > \omega_1$)

Example: demodulation

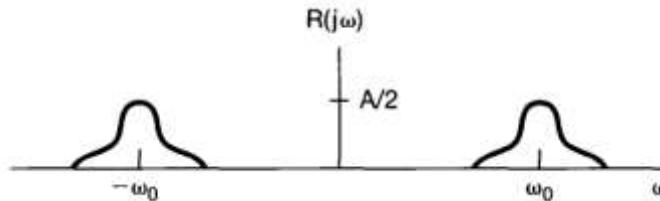
take the same $r(t)$

multiply again by $p(t)$

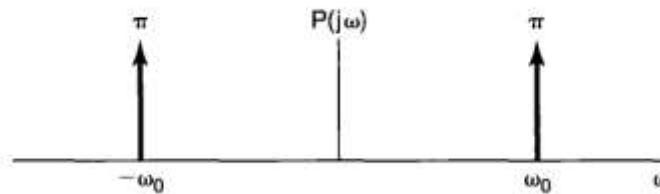


Example: demodulation

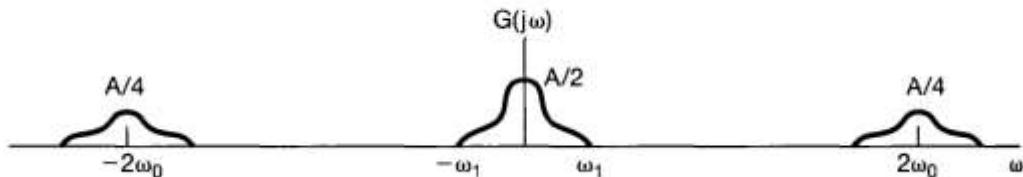
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$$g(t) = r(t) \cdot p(t)$$

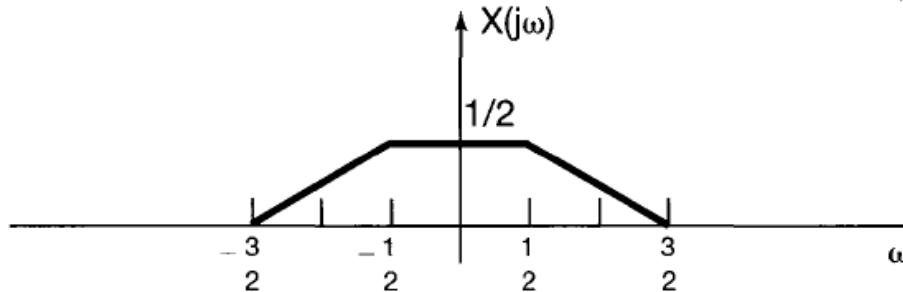


we can recover $s(t)$ with a lowpass filter

Example: multiplication property

$$x(t) = \frac{\sin(t) \sin(t/2)}{\pi t^2} = \pi \left(\frac{\sin(t)}{\pi t} \right) \left(\frac{\sin(t/2)}{\pi t} \right)$$

$$X(j\omega) = \frac{1}{2} \mathcal{F} \left\{ \frac{\sin(t)}{\pi t} \right\} * \mathcal{F} \left\{ \frac{\sin(t/2)}{\pi t} \right\}$$



Systems characterized by linear constant-coefficient differential equations

$$\sum_{k=0}^N a_k \frac{d^k y(t)}{dt^k} = \sum_{k=0}^M b_k \frac{d^k x(t)}{dt^k}$$

To determine the frequency response $H(j\omega)$
we Fourier transform both sides

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$$\sum_{k=0}^N a_k (j\omega)^k Y(j\omega) = \sum_{k=0}^M b_k (j\omega)^k X(j\omega)$$

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$$\sum_{k=0}^N a_k (j\omega)^k Y(j\omega) = \sum_{k=0}^M b_k (j\omega)^k X(j\omega)$$

$$H(j\omega) = \frac{Y(j\omega)}{X(j\omega)} = \frac{\sum_{k=0}^M b_k (j\omega)^k}{\sum_{k=0}^N a_k (j\omega)^k}$$

ratio of polynomials

Examples

Example: first-order system

$$\frac{dy(t)}{dt} + a y(t) = x(t) \quad a > 0$$

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$$H(j\omega) = \frac{Y(j\omega)}{X(j\omega)} = \frac{1}{j\omega + a} \quad \xleftrightarrow{\mathcal{F}} \quad h(t) = e^{-at} u(t)$$

Example: second-order system

$$\frac{d^2 y(t)}{dt^2} + 4 \frac{dy(t)}{dt} + 3y(t) = \frac{dx(t)}{dt} + 2x(t)$$

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$$H(j\omega) = \frac{j\omega + 2}{(j\omega)^2 + 4j\omega + 3} = \frac{j\omega + 2}{(j\omega + 1)(j\omega + 3)} = \frac{1/2}{j\omega + 1} + \frac{1/2}{j\omega + 3}$$

$$h(t) = \frac{1}{2}e^{-t} u(t) + \frac{1}{2}e^{-3t} u(t)$$

now assume $x(t) = e^{-t}u(t)$, what is $y(t)$?

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$$\begin{aligned} Y(j\omega) &= H(j\omega) \cdot X(j\omega) = \frac{j\omega + 2}{(j\omega + 1)(j\omega + 3)} \cdot \frac{1}{j\omega + 1} \\ &= \frac{j\omega + 2}{(j\omega + 1)^2 (j\omega + 3)} = \dots = \frac{1/4}{j\omega + 1} + \frac{1/2}{(j\omega + 1)^2} - \frac{1/4}{j\omega + 3} \end{aligned}$$

$$y(t) = \frac{1}{4} e^{-t} u(t) + \frac{1}{2} t e^{-t} u(t) - \frac{1}{4} e^{-3t} u(t)$$

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