

# Signals and Systems

## #07: Convolution property

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*Some of the images in this presentation are from “Signals and Systems”, A. V. Oppenheim, A. S. Wilsky, and S. H. Nawab, 2<sup>nd</sup> ed. Pearson.*

# Outline

- #01 Continuous-time signals
- #02 Continuous-time systems
- #03 Fourier series of periodic signals
- #04 Fourier transform of aperiodic signals
- #05 Fourier transform of periodic signals
- #06 Basic properties of the Fourier transform
- **#07 Convolution property**
- #08 Multiplication property

# Convolution property

Recall  $h(t) * x(t) \triangleq \int_{-\infty}^{\infty} x(\tau)h(t - \tau)d\tau$

We have

$$y(t) = h(t) * x(t) \xleftrightarrow{\mathcal{F}} Y(j\omega) = H(j\omega) \cdot X(j\omega)$$

$H(j\omega) = \mathcal{F} \{h(t)\}$  is the **frequency response**

Recall  $h(t) * x(t) \triangleq \int_{-\infty}^{\infty} x(\tau)h(t - \tau)d\tau$

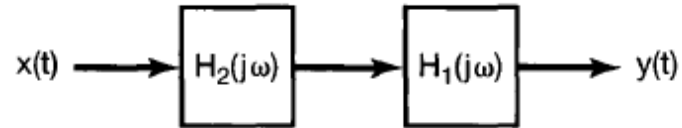
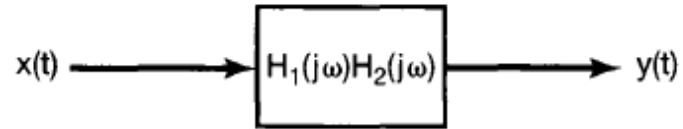
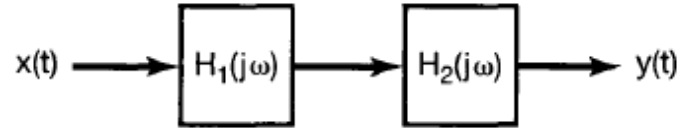
We have

$$y(t) = h(t) * x(t) \xleftrightarrow{\mathcal{F}} Y(j\omega) = H(j\omega) \cdot X(j\omega)$$

[Proof:]

$$\begin{aligned} Y(j\omega) &= \mathcal{F}\{y(t)\} = \int_{-\infty}^{\infty} \left[ \int_{-\infty}^{\infty} x(\tau)h(t - \tau)d\tau \right] e^{-j\omega t} dt \\ &= \int_{-\infty}^{\infty} x(\tau) \left[ \int_{-\infty}^{\infty} h(t - \tau)e^{-j\omega t} dt \right] d\tau \\ &= \int_{-\infty}^{\infty} x(\tau) H(j\omega) e^{-j\omega\tau} d\tau = H(j\omega) \cdot X(j\omega) \end{aligned}$$

## Cascade of systems



# Examples

Example: time-shift system

$$h(t) = \delta(t - t_0) \xleftrightarrow{\mathcal{F}} H(j\omega) = e^{-j\omega t_0}$$

For an input  $x(t)$  with  $\mathcal{F}\{x(t)\} = X(j\omega)$

the output is

$$y(t) = x(t) * \delta(t - t_0) = x(t - t_0) \quad \text{with}$$

$$\mathcal{F}\{y(t)\} = Y(j\omega) = H(j\omega) \cdot X(j\omega) = X(j\omega) e^{-j\omega t_0}$$



Example: differentiator

$$y(t) = \frac{dx(t)}{dt}$$

therefore

$$Y(j\omega) = j\omega X(j\omega) \quad \text{and}$$

$$H(j\omega) = j\omega$$

Example: integrator

$$\begin{aligned}y(t) &= \int_{-\infty}^t x(\tau) d\tau \\ &= x(t) * u(t)\end{aligned}$$

therefore

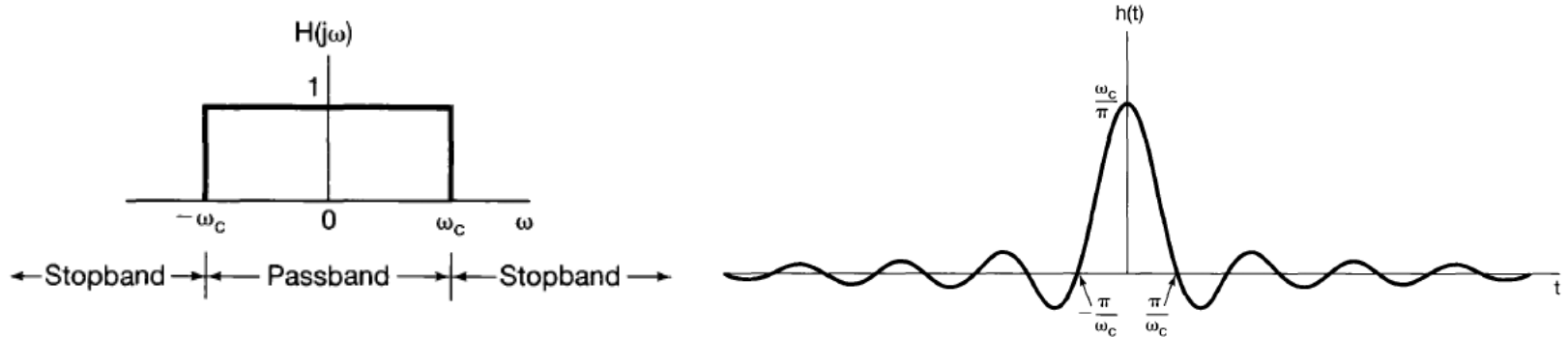
$$H(j\omega) = \mathcal{F}\{u(t)\} = \frac{1}{j\omega} + \pi \delta(\omega) \quad \text{and}$$

$$Y(j\omega) = H(j\omega) \cdot X(j\omega) = \frac{1}{j\omega} X(j\omega) + \pi X(0) \delta(\omega)$$

Example: frequency-selective filter (ideal)

$$H(j\omega) = \begin{cases} 1 & |\omega| < \omega_c \\ 0 & |\omega| > \omega_c \end{cases} \quad \xleftrightarrow{\mathcal{F}} \quad h(t) = \frac{\sin \omega_c t}{\pi t}$$

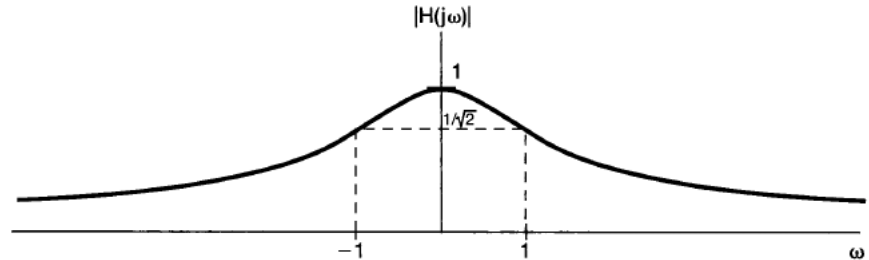
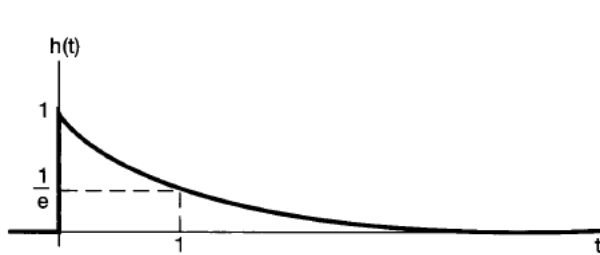
perfect frequency selectivity  $\Rightarrow$  non-causal filter with oscillations



Example: frequency-selective filter (smooth)

$$h(t) = e^{-t} u(t) \quad \xleftrightarrow{\mathcal{F}} \quad H(j\omega) = \frac{1}{j\omega + 1}$$

causal without oscillations  $\Rightarrow$  weaker frequency selectivity



Example: evaluating convolution integral

$$h(t) = e^{-at} u(t), \quad a > 0$$

$$x(t) = e^{-bt} u(t), \quad b > 0$$

$$y(t) = h(t) * x(t) ?$$

$$H(j\omega) = \frac{1}{a + j\omega}$$

$$X(j\omega) = \frac{1}{b + j\omega}$$

$$Y(j\omega) = H(j\omega) \cdot X(j\omega) = \frac{1}{(a + j\omega)(b + j\omega)} = \dots$$

For  $b \neq a$

$$Y(j\omega) = \frac{1}{(a + j\omega)(b + j\omega)} = \frac{A}{a + j\omega} + \frac{B}{b + j\omega} = \dots$$

$$Ab + Aj\omega + Ba + Bj\omega = 1 \quad \Rightarrow \quad \begin{cases} A = \frac{1}{b-a} \\ B = \frac{1}{a-b} \end{cases}$$

$$\dots = \frac{1}{b-a} \left[ \frac{1}{a + j\omega} - \frac{1}{b + j\omega} \right] \quad \xleftrightarrow{\mathcal{F}} \quad y(t) = \frac{1}{b-a} [e^{-at} - e^{-bt}] u(t)$$

For  $b = a$

$$Y(j\omega) = \frac{1}{(a + j\omega)^2} = \dots$$

recall

$$e^{-at} u(t) \xleftrightarrow{\mathcal{F}} \frac{1}{a + j\omega}$$
$$t e^{-at} u(t) \xleftrightarrow{\mathcal{F}} j \frac{d}{d\omega} \left[ \frac{1}{a + j\omega} \right] = \frac{1}{(a + j\omega)^2}$$

$$\dots = j \frac{d}{d\omega} \left[ \frac{1}{a + j\omega} \right] \xleftrightarrow{\mathcal{F}} y(t) = t e^{-at} u(t)$$

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