## Signals and Systems 2

Giovanni Geraci — Universitat Pompeu Fabra, Barcelona

Tutorial #4: Sampling/reconstruction and modulation/demodulation

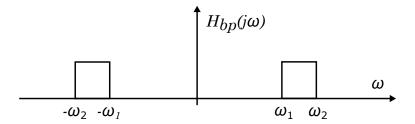
- 1. Consider two signals  $x_1(t)$  and  $x_2(t)$ , bandlimited so that  $X_1(j\omega) = 0, |\omega| \ge \omega_1$  and  $X_2(j\omega) = 0, |\omega| \ge \omega_2$ . Both signals are multiplied, so that  $w(t) = x_1(t) \cdot x_2(t)$ , and then sampled by a periodic impulse train  $p(t) = \sum_{n=-\infty}^{+\infty} \delta(t-n \cdot T_s)$ . Determine the maximum sampling period  $T_s$  so that w(t) can be perfectly reconstructed by an ideal LPF.
- 2. Determine the Nyquist rate of the signals below:
  - (a)  $x(t) = 1 + \cos(2000\pi t) + \sin(4000\pi t)$ (b)  $x(t) = \frac{\sin(4000\pi t)}{\pi t}$ (c)  $x(t) = (\frac{\sin(4000\pi t)}{\pi t})^2$
- 3. Consider a signal x(t) such that its Nyquist rate is  $\omega_0$ . Determine the Nyquist rate of the signals below:
  - (a) x(t) + x(t-1)(b)  $x^{2}(t)$ (c)  $x(t) \cdot \cos(\omega_{0}t)$
- 4. Consider a signal x(t) such that  $X(j\omega) = 0, |\omega| \ge \omega_b$ . The signal is modulated by multiplication with a carrier  $p(t) = \cos(\omega_p t)$ , resulting in  $x_m(t)$ . If we sample  $x_m(t)$  with an impulse train of period  $T_s$ , determine the maximal value of  $\omega_p$  so that we can perfectly reconstruct  $x_m(t)$ .

- 5. Determine whether the following statements are true or false:
  - (a) The signal  $x(t) = u(t + T_0) u(t T_0)$  can be sampled correctly (without aliasing) with an impulse train if the sampling period is  $T_s < 2T_0$ .
  - (b) The signal with Fourier transform  $X(j\omega) = u(\omega + \omega_0) u(\omega \omega_0)$  can be sampled correctly (without aliasing) with an impulse train if the sampling period is  $T_s < \frac{\pi}{\omega_0}$ .
  - (c) The signal with Fourier transform  $X(j\omega) = u(\omega) u(\omega \omega_0)$  can be sampled correctly (without aliasing) with an impulse train if the sampling period is  $T_s < \frac{2\pi}{\omega_0}$ .
- 6. Consider a signal x(t) satisfying  $X(j\omega) = \begin{cases} \neq 0, & \omega_1 \le |\omega| \le \omega_2 \\ 0, & \text{otherwise} \end{cases}$ , where  $\omega_1 > \omega_2 \omega_1$ . Such signals, named *passband signals*, can be perfectly reconstructed after sampling with an impulse train with sampling frequency  $\omega_s$  below the Nyquist rate, by using a recontent of A,  $\omega_a \le |\omega| \le \omega_b$

struction filter 
$$H(j\omega) = \begin{cases} A, & \omega_a \ge |\omega| \ge 0\\ 0, & \text{otherwise} \end{cases}$$

Determine  $\omega_a, \omega_b$ , and the range of  $\omega_s$  so that perfect reconstruction is possible.

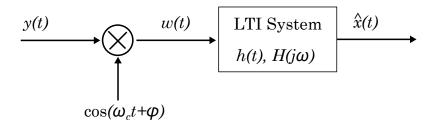
7. Consider the ideal bandpass filter (with unitary amplitude) shown below:



Show that its impulse response,  $h_{bp}(t)$ , can be represented as any of the following three ways:

- (a) A cosine-modulated lowpass filter:  $h_{bp}(t) = A \frac{\sin(\omega_{co}t)}{\pi t} \cos(\omega_0 t)$  (Note: determine A,  $\omega_{co}$ , and  $\omega_0$ ).
- (b) The difference of two ideal lowpass filters:  $h_{bp}(t) = \frac{\sin(\omega_2 t)}{\pi t} \frac{\sin(\omega_1 t)}{\pi t}$ .
- (c) Convolution of the impulse responses of an ideal highpass filter and an ideal lowpass filter:  $h_{bp}(t) = \left[\delta(t) \frac{\sin(\omega_1 t)}{\pi t}\right] * \frac{\sin(\omega_2 t)}{\pi t}$ .

8. Consider the demodulation system below:

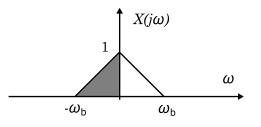


where

$$H(j\omega) = \begin{cases} 2, & |\omega| \le \omega_{co} \\ 0, & \text{otherwise} \end{cases}$$
(1)

We have shown that when  $\phi = 0$ , if we apply this system to the modulated signal  $y(t) = x(t)\cos(\omega_c t)$ , we can recover the original signal x(t) (i.e.,  $\hat{x}(t) = x(t)$ ), provided that x(t) is bandlimited and the frequencies  $\omega_{co}$  and  $\omega_c$  are properly chosen.

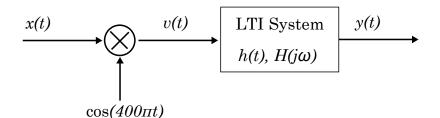
Assume that x(t) has the following bandlimited Fourier transform representation:



and that  $\omega_c > \omega_b$ .

- (a) For  $\phi \neq 0$ , use Euler's formula to show that:  $w(t) = \frac{1}{2}x(t)\cos(\phi) + \frac{1}{2}x(t)\cos(2\omega_c t + \phi)$ .
- (b) Use the result in part (a) to obtain  $W(j\omega)$  and plot it.
- (c) Assuming  $\omega_{co} = \omega_b$ , plot  $\hat{X}(j\omega)$ .
- (d) Obtain an equation for  $\hat{x}(t)$  in terms of x(t) and  $\phi$ .

9. Consider the amplitude modulation system below:



and the bandlmited input signal x(t) of exercise 8 with  $\omega_b = 100\pi$ . Assume that the LTI system has the frequency response of an ideal bandpass filter:

$$H(j\omega) = \begin{cases} 1, & 300\pi < |\omega| \le 400\pi\\ 0, & \text{otherwise} \end{cases}$$
(2)

- (a) Plot  $H(j\omega)$  and  $V(j\omega)$ .
- (b) Use the plots in part (a) to plot  $Y(j\omega)$ . Note that the negative frequency portion of  $X(j\omega)$  is shaded. Mark the corresponding regions in your plot of  $V(j\omega)$  and  $Y(j\omega)$ .
- (c) The output signal is called a "single-sideband" signal. Justify this terminology.
- (d) How could you recover x(t) from the single-sideband signal y(t)? Draw a block diagram of the system that recovers the original signal.
- 10. Consider the signal x(t) as in exercise 8, with  $\omega_b = 80\pi$ . We sample the signal x(t) at frequency  $\omega_s$ , and then we reconstruct it with a filter that has the following frequency response:

$$H_r(j\omega) = \begin{cases} T, & |\omega| \le \pi/T_s \\ 0, & \text{otherwise} \end{cases}$$
(3)

- (a) Choose the value of  $\omega_s$  so that it is equal to the Nyquist rate and plot  $X_s(j\omega)$ .
- (b) If  $\omega_s = 2\pi/T_s = 100\pi$ , plot  $X_s(j\omega)$  and use that to plot  $X_r(j\omega)$ . What do you observe?